



Vibration energy harvesting systems

Francesco Cottone

NiPS laboratory, Department of Physics and Geology,
Università di Perugia, Italy

ERASMUS + IESRES
INNOVATIVE EUROPEAN STUDIES on RENEWABLE ENERGY SYSTEMS
Teaching Activity
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Outline

- Energy harvesting applications and principles
- Fundamentals of vibration energy harvesters
- Beyond linear systems
- Microscale energy harvesters
- Final considerations

Energy harvesting applications

Structural Monitoring

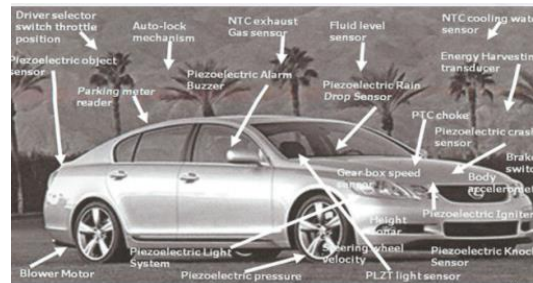


02/07/2014 - Belo Horizonte (Brazil)
(birdge collapse at FIAT factory)

Environmental Monitoring

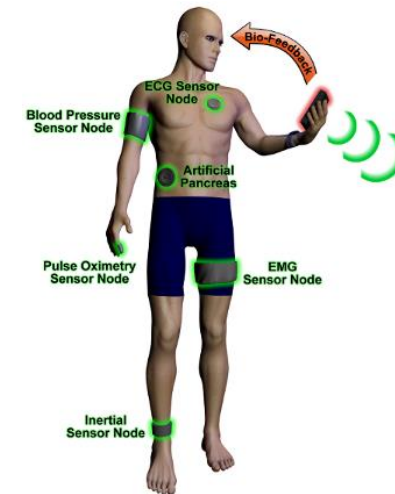


Transportation



Wearable sensing for health applications

Emergency medical response
Monitoring, pacemaker, defibrillators



Military applications



Wireless Sensor Networks

Energy Harvesting could enable 90% of WSNs applications (IdTechex)

Power sources available from the ambient

Thermal energy

Radioactivity

EM energy

Solar

RF



Vibrations

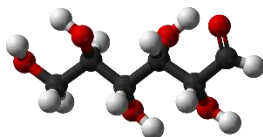
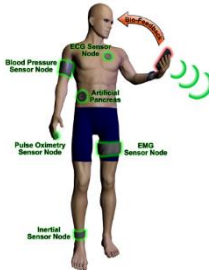


Hydro/wind



Traffic

Biochemical



Energy
Harvesting
Generator

Temporary
Storage
system

Electronic
device

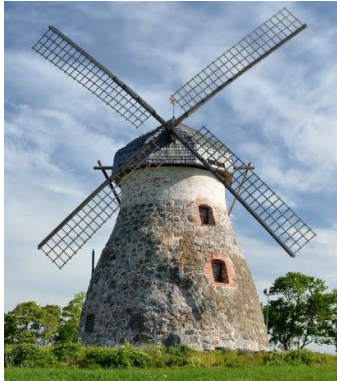
Wasted thermal energy

- Piezoelectric
- Electrodynamics
- Photovoltaic
- Hydro Turbine

- Ultra capacitors
- Rechargeable Batteries

- Wireless Sensor Node

Human-made energy harvesters



Wind mill (Origin: Persia, 3000 years BC)



Sailing ship (XVI-XVII century)



Crystal radio - 1906

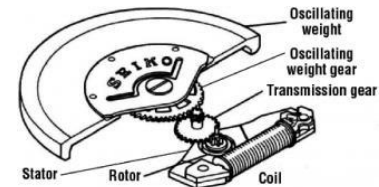


SELF-powered by Radio Frequencies !!!



First automatic wristwatch, Harwood, c. 1929 (Deutsches Uhrenmuseum, Inv. 47-3543)

First automatic watch.
Abraham-Louis Perrelet,
Le Locle. 1776



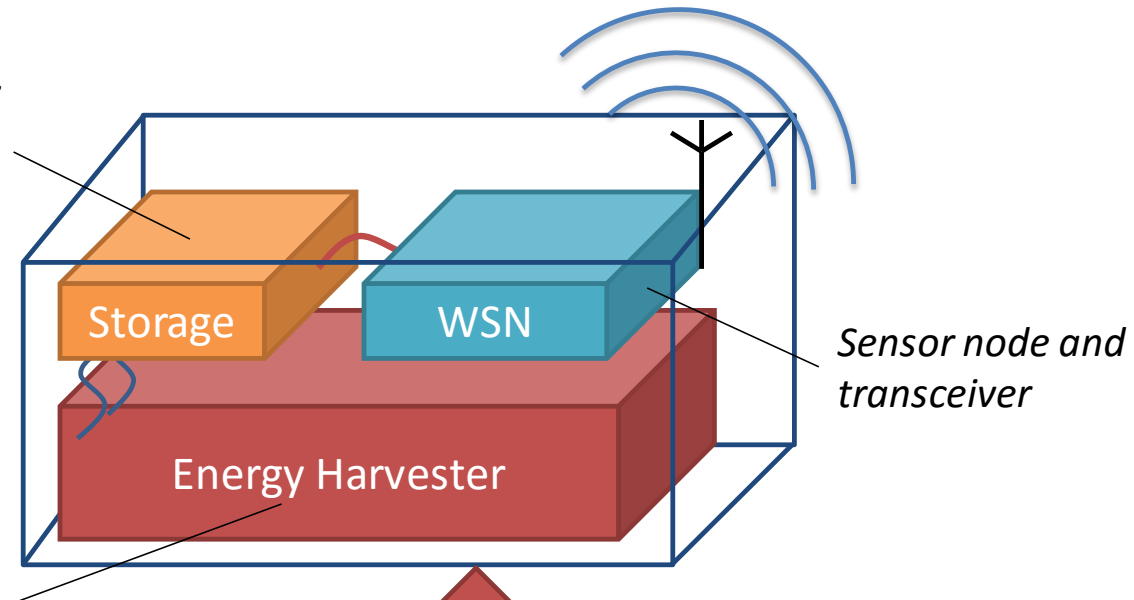
Self-charging Seiko
wristwatch

Vibration powered wireless sensor

Power density objective for EH:
 $> 100 \mu\text{W}/\text{cm}^3$

*Temporary storage
and conditioning electronics:*

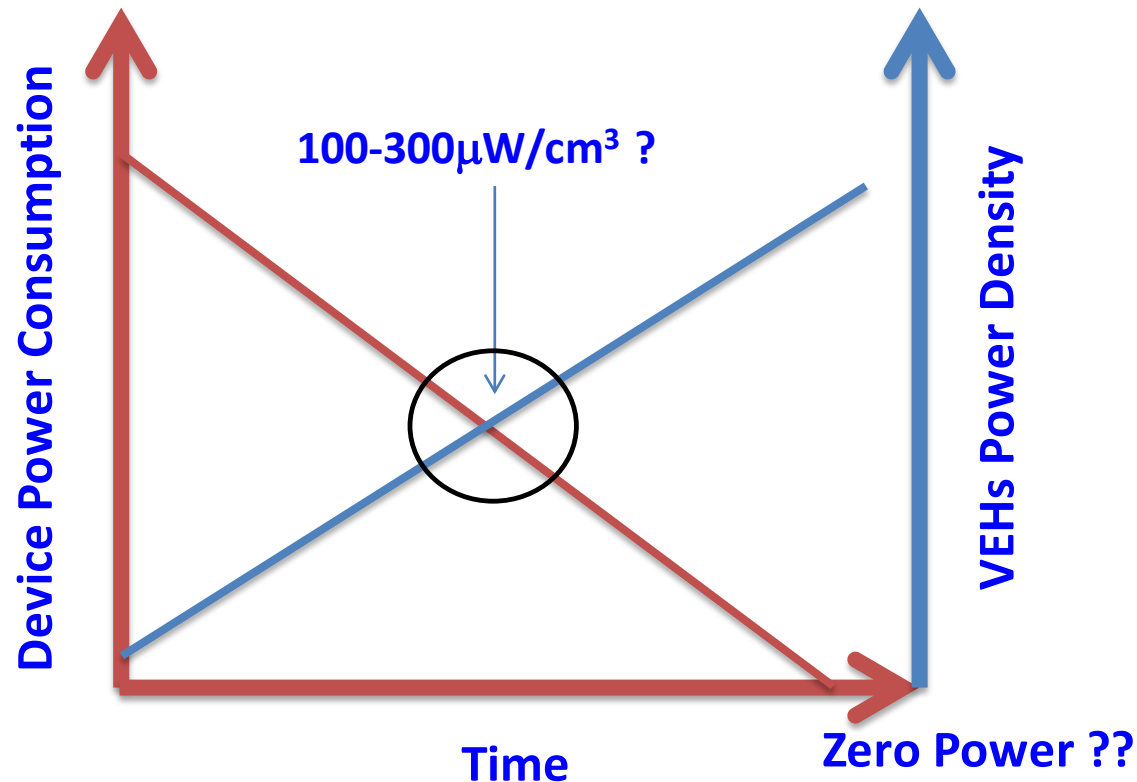
- Ultra capacitors
- Rechargeable Batteries



Energy harvesting system:

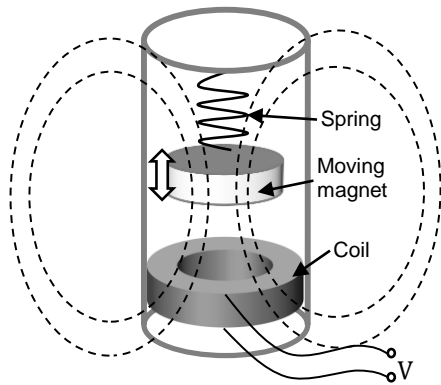
- piezoelectric,
- electromagnetic,
- electrostatic,
- magnetostrictive

Mobile devices: power needs

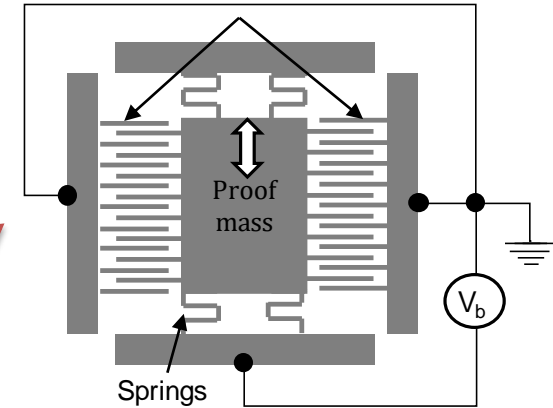


Vibration energy harvesting

Electromagnetic

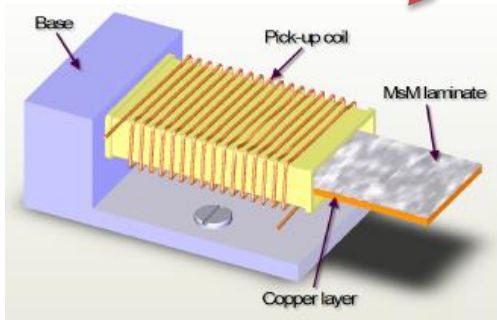


Electrostatic/Capacitive

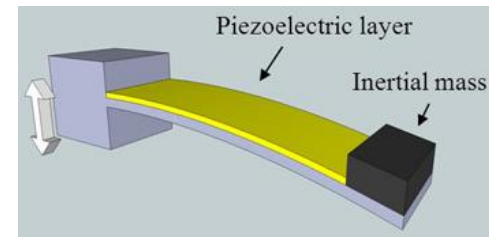


Vibration Harvesting Generator

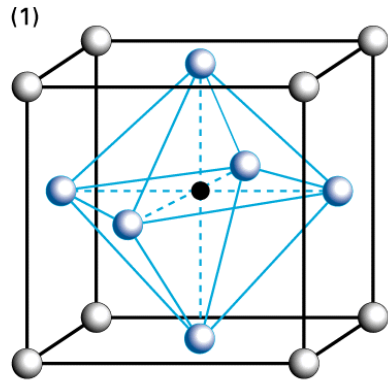
Magnetostrictive



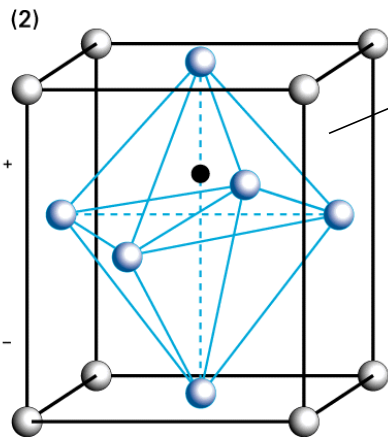
Piezoelectric



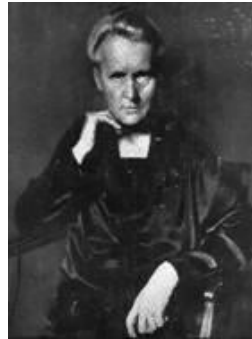
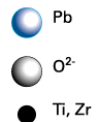
Piezoelectric conversion



Unpolarized
Crystal

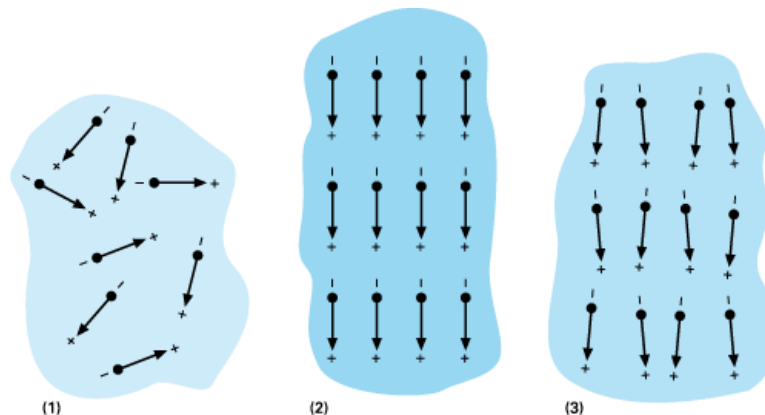


Polarized
Crystal



Pioneering work on the direct piezoelectric effect (stress-charge) in this material was presented by **Jacques and Pierre Curie in 1880**

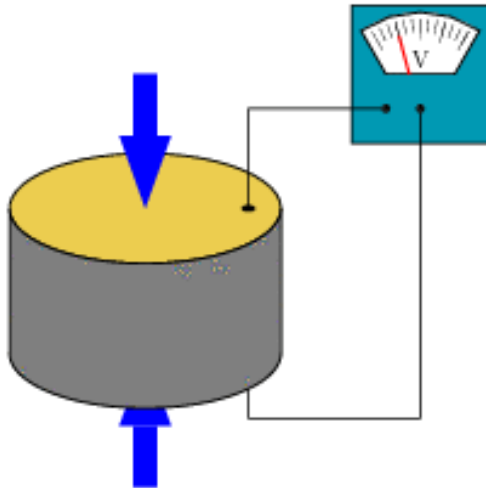
In 1903 Pierre received the Nobel Prize in Physics with his wife, Marie Skłodowska-Curie and Henri Becquerel, for the research on the radiation phenomena discovered by Professor Henri Becquerel.



After poling the zirconate-titanate atoms are off center. The molecule becomes elongated and polarized

Piezoelectric conversion

Stress-to-charge conversion



direct piezoelectric effect

Biological

- Bones
- [DNA !!!](#)

Naturally-occurring crystals

- [Berlinite](#) (AlPO_4), a rare [phosphate mineral](#) that is structurally identical to quartz
- [Cane sugar](#)
- [Quartz](#) (SiO_2)
- [Rochelle salt](#)

Man-made ceramics

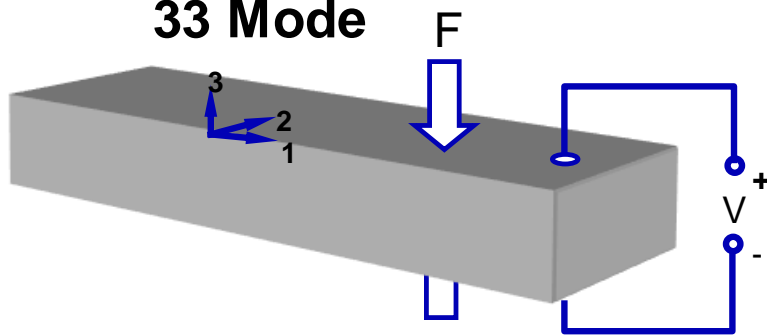
- [Barium titanate](#) (BaTiO_3)—Barium titanate was the first piezoelectric ceramic discovered.
- [Lead titanate](#) (PbTiO_3)
- [Lead zirconate titanate](#) ($\text{Pb}[\text{Zr}_x\text{Ti}_{1-x}]\text{O}_3$ $0 \leq x \leq 1$)—more commonly known as **PZT**, lead zirconate titanate is the most common piezoelectric ceramic in use today.
- [Lithium niobate](#) (LiNbO_3)

Polymers

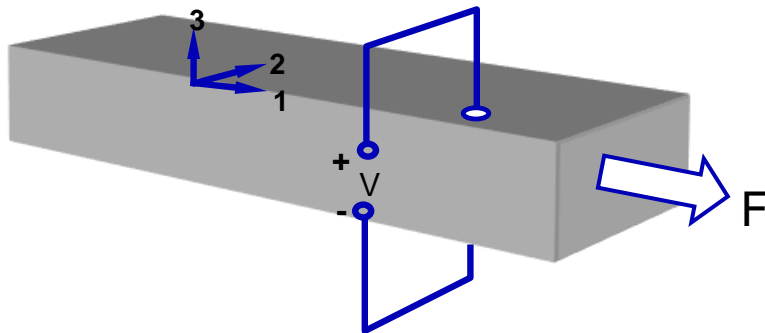
- [Polyvinylidene fluoride](#) (PVDF): exhibits piezoelectricity several times greater than quartz. Unlike ceramics, long-chain molecules attract and repel each other when an electric field is applied.

Piezoelectric conversion

33 Mode



31 Mode



$$S = [s_E]T + [d^t]E$$

Strain-charge

$$D = [d]T + [\varepsilon_T]E$$

$$T = [c^E]S - [e^t]E$$

Stress-charge

$$D = [e]S + [\varepsilon^S]E$$

- S = strain vector (6x1) in Voigt notation
- T = stress vector (6x1) [N/m²]
- s_E = compliance matrix (6x6) [m²/N]
- c^E = stiffness matrix (6x6) [N/m²]
- d = piezoelectric coupling matrix (3x6) in Strain-Charge [C/N]
- D = electrical displacement (3x1) [C/m²]
- e = piezoelectric coupling matrix (3x6) in Stress-Charge [C/m²]
- ε = electric permittivity (3x3) [F/m]
- E = electric field vector (3x1) [N/C] or [V/m]

Piezoelectric conversion

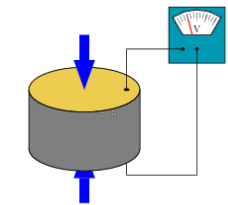
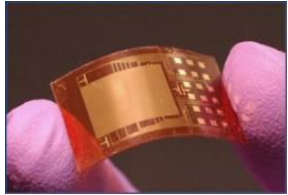
Characteristic	PZT-5H	BaTiO3	PVDF	AlN (thin film)
d_{33} (10^{-10} C/N)	593	149	-33	5,1
d_{31} (10^{-10} C/N)	-274	78	23	-3,41
k_{33}	0,75	0,48	0,15	0,3
k_{31}	0,39	0,21	0,12	0,23
ϵ_r	3400	1700	12	10,5

$$k_{31}^2 = \frac{\text{El.energy}}{\text{Mech.energy}} = \frac{d_{31}^2}{s_{11}^E \epsilon_{33}^T}$$

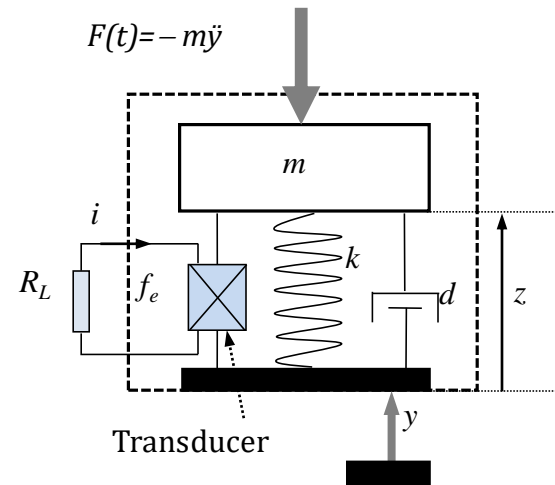
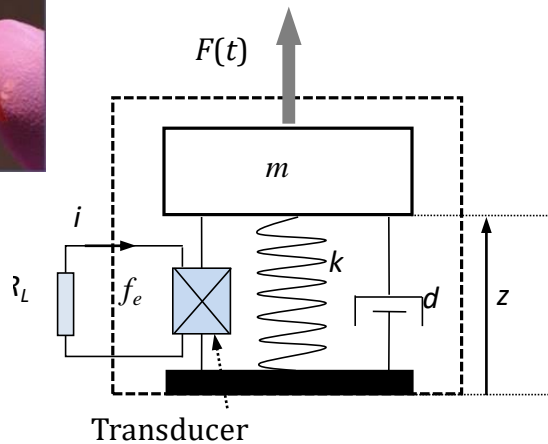
Electromechanical Coupling is an adimensional factor that provides the effectiveness of a piezoelectric material. IT's defined as the ratio between the mechanical energy converted and the electric energy input or the electric energy converted per mechanical energy input

Basic model of VEH

zinc oxide (ZnO) nanowires
Wang et al. 2008

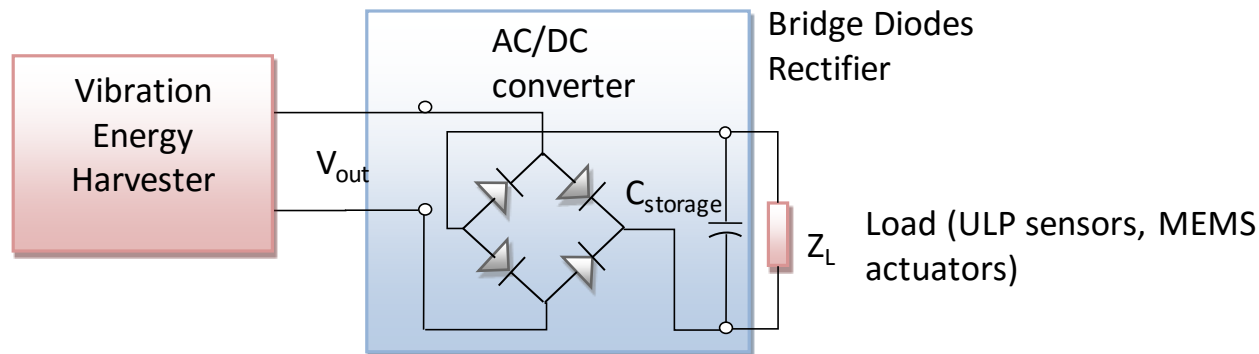


Energy Harvesting from dancing

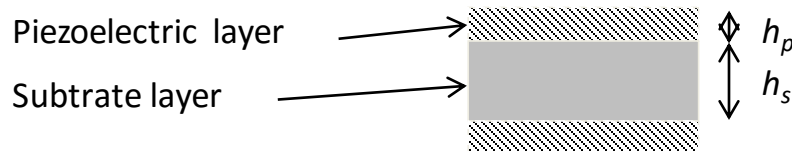
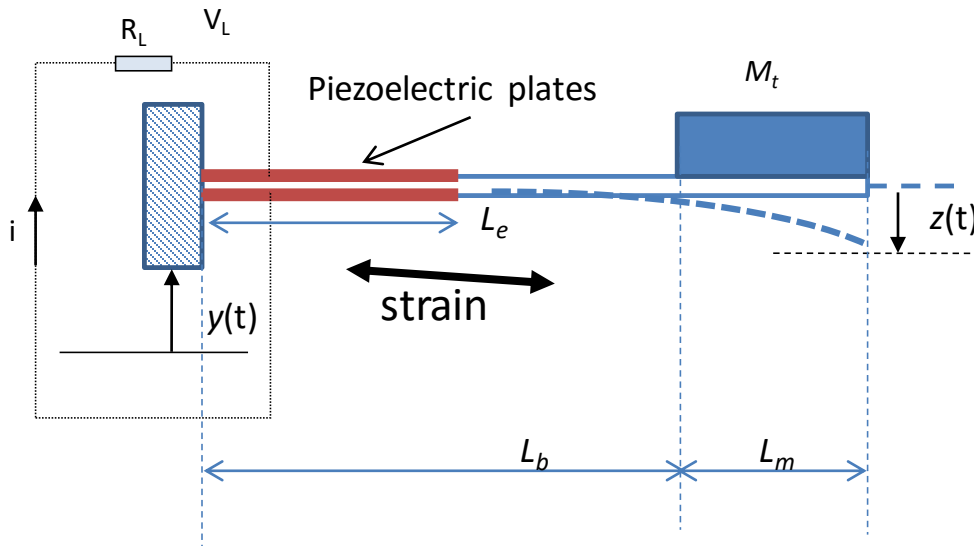


Energy harvesting from
moth vibrations
Chang, MIT 2013

Inertial generators require only one point of attachment to a moving structure, allowing a greater degree of miniaturization.



Piezoelectric conversion



E_p and E_s are the Young's modulus of piezo layer and steel substrate respectively

Governing equations

$$\begin{cases} m\ddot{z} + d\dot{z} + kz + \alpha V_L = -m\ddot{y} \\ \dot{V}_L + (\omega_c + \omega_i)V_L = \lambda\omega_c \dot{z} \end{cases}$$



$$\alpha = kd_{31} / h_p k_2,$$

$$\lambda = \alpha R_L,$$

$$\omega_c = 1 / R_L C_p,$$

$$\omega_i = 1 / R_i C_p,$$

$$k = k_1 k_2 E_p,$$

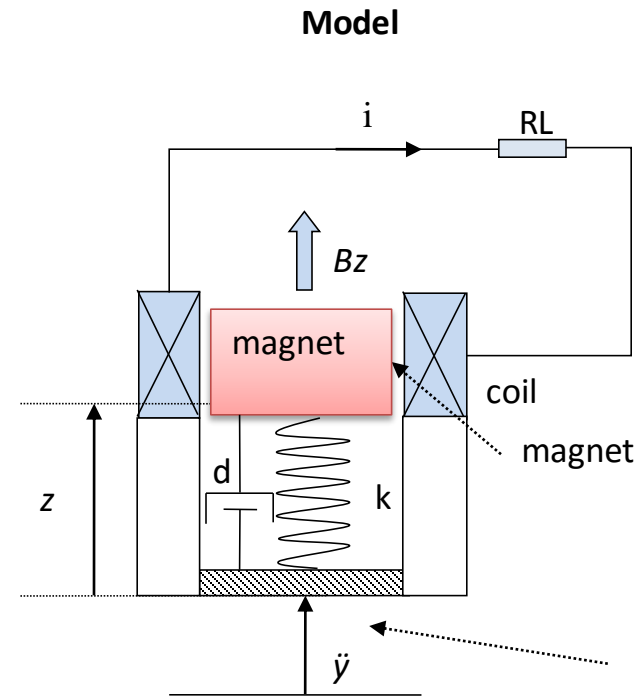
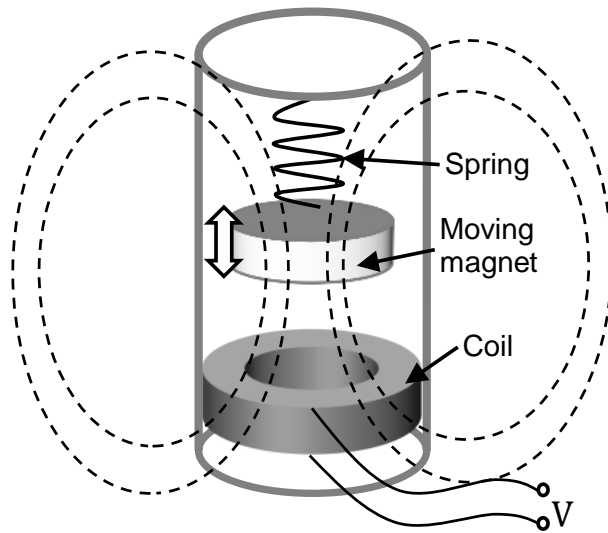
$$k_1 = \frac{2I}{b(2l_b + l_m - l_e)},$$

$$k_2 = \frac{3b(2l_b + l_m - l_e)}{l_b^2 \left(2l_b + \frac{3}{2}l_m \right)},$$

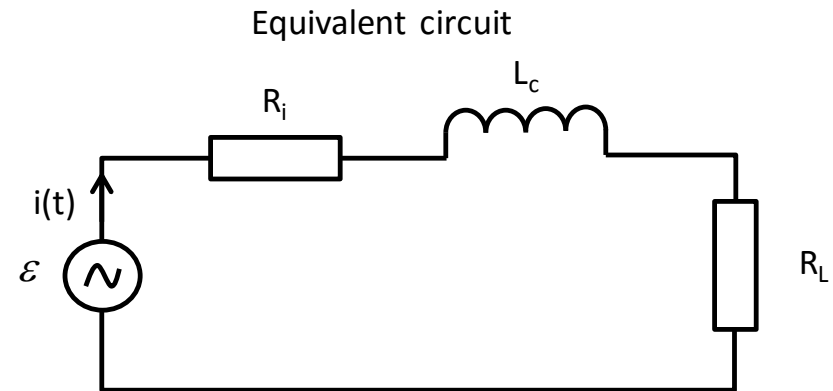
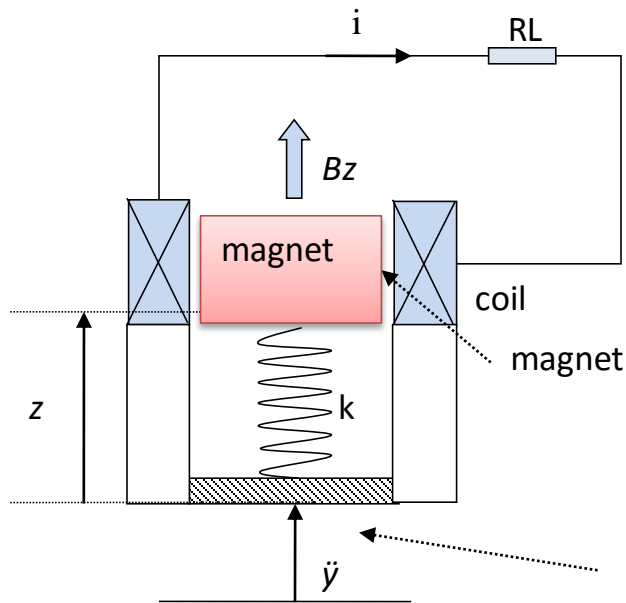
$$b = \frac{h_s + h_p}{2},$$

$$I = 2 \left[\frac{w_b h_p^3}{12} + w_b h_p b^2 \right] + \frac{E_s / E_p w_b h_s^3}{12},$$

Electromagnetic conversion



Electromagnetic conversion



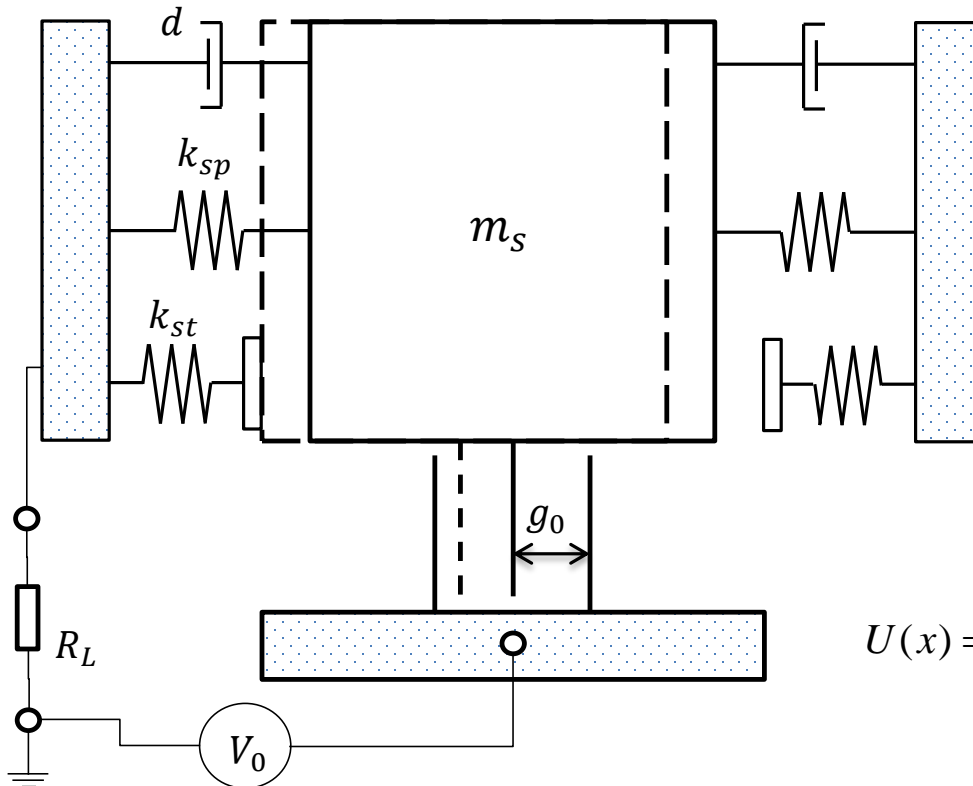
Governing equations

$$\begin{cases} m\ddot{z} + d\dot{z} + kz + \alpha V_L = -m\ddot{y} \\ \dot{V}_L + (\omega_c + \omega_i)V_L = \lambda\omega_c\dot{z} \end{cases}$$



$$\begin{aligned} \alpha &= Bl / R_L, & \lambda &= Bl = \alpha R_L, \\ \omega_c &= R_L / L_c, & \omega_i &= R_i / L_c, \end{aligned}$$

Electrostatic conversion



Governing equations

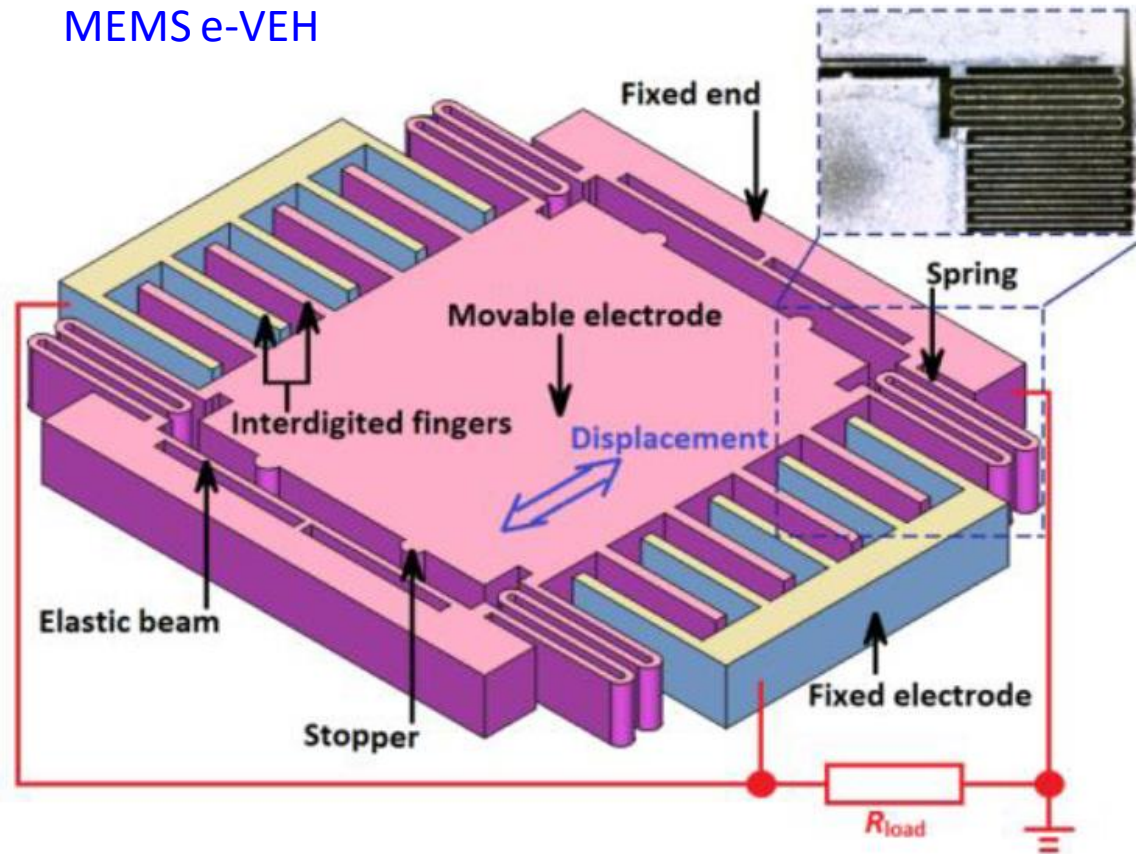
$$m \frac{d^2 x}{dt^2} + (c_a + c_i) \frac{dx}{dt} + \frac{dU(x)}{dx} = -m \frac{d^2 y}{dt^2},$$

$$R_L \frac{d}{dt} (C \cdot V) + V = U_0,$$

$$U(x) = \begin{cases} \frac{1}{2} k_{sp} x^2 - \frac{1}{2} C(x) U_0^2, & \text{for } |x| < x_{\text{lim}} \\ \frac{1}{2} (k_{sp} + k_{st}) x^2 - \frac{1}{2} C(x) U_0^2, & \text{for } |x| \geq x_{\text{lim}} \end{cases}$$

Electrostatic conversion

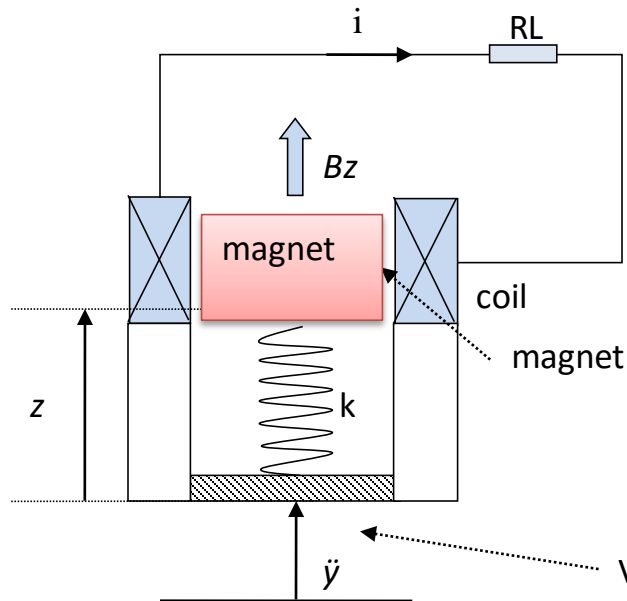
MEMS e-VEH



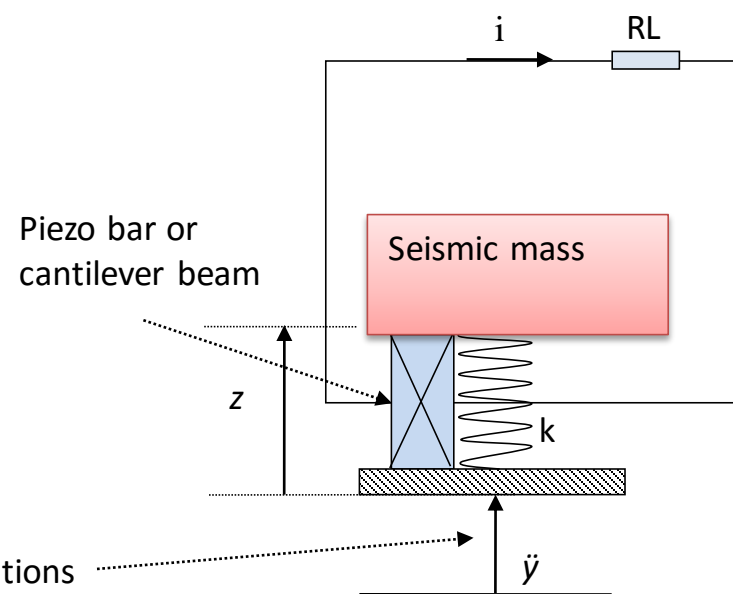
Y. Lu, F. Cottone, S. Boisseau, F. Marty, D. Galayko, and P. Basset, *Applied Physics Letters* 2015.

General model

Electromagnetic transduction



Piezoelectric transduction



$$\begin{cases} m\ddot{z} + d\dot{z} + \frac{dU(z)}{dz} + \alpha V_L = -m\ddot{y} \\ \dot{V}_L + (\omega_c + \omega_i)V_L = \lambda\omega_c\dot{z} \end{cases}$$

$$\Rightarrow \alpha, \lambda, \omega_c, \omega_i$$

Depends on the specific transduction technique!

General model

For LINEAR mechanical oscillators

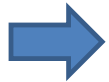


$$\begin{cases} m\ddot{z} + d\dot{z} + kz + \alpha V_L = -m\ddot{y} \\ \dot{V}_L + (\omega_c + \omega_i)V_L = \lambda\omega_c \dot{z} \end{cases}$$

Laplace transform

$$\ddot{y} = Y_0 e^{j\omega t} \quad \Rightarrow \quad \begin{pmatrix} ms^2 + ds + k & \alpha \\ -\lambda\omega_c s & s + \omega_c \end{pmatrix} \begin{pmatrix} Z \\ V \end{pmatrix} = \begin{pmatrix} -mY \\ 0 \end{pmatrix}$$

$$Z = \frac{-mY}{\det A} (s + \omega_c) = \frac{-mY \cdot (s + \omega_c)}{ms^3 + (m\omega_c + d)s^2 + (k + \alpha\lambda\omega_c + d\omega_c)s + k\omega_c},$$



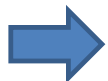
$$V = \frac{-mY}{\det A} \lambda\omega_c s = \frac{-mY \cdot \lambda\omega_c s}{ms^3 + (m\omega_c + d)s^2 + (k + \alpha\lambda\omega_c + d\omega_c)s + k\omega_c}.$$

Hence, the transfer functions between displacement and voltage over input acceleration are given by

$$H_{zy}(s) = \frac{Z}{Y}, \quad (a)$$



$$H_{vy}(s) = \frac{V}{Y}. \quad (b)$$

By substituting $s=j\omega$ in , we can calculate the electrical power dissipated across the resistive load



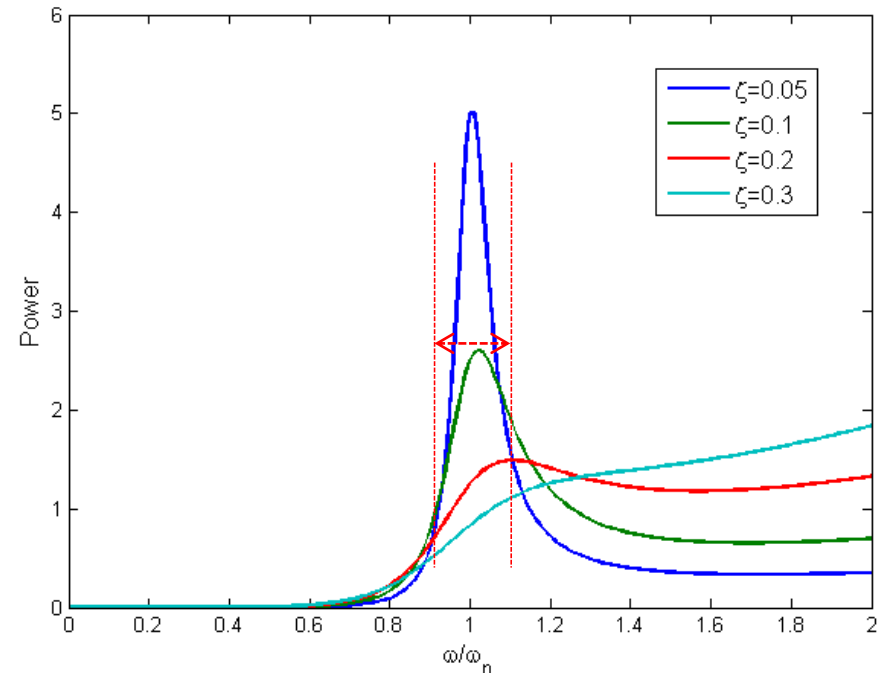
$$P_e(\omega) = \frac{Y_0^2}{2R_L} \left| \frac{m_2 \lambda \omega_c j\omega}{(\omega_c + j\omega)(-m_2 \omega^2 + d_2 j\omega + k_2) + \alpha \lambda \omega_c j\omega} \right|^2$$

Comparison of conversion techniques

Technique	Advantages 	Drawbacks 
Piezoelectric	<ul style="list-style-type: none"> • high output voltages • well adapted for miniaturization • high coupling in single crystal • no external voltage source needed 	<ul style="list-style-type: none"> • expensive • small coupling for piezoelectric thin films • large load optimal impedance required ($M\Omega$) • Fatigue effect
Electrostatic	<ul style="list-style-type: none"> • suited for MEMS integration • good output voltage (2-10V) • possibility of tuning electromechanical coupling • Long-lasting 	<ul style="list-style-type: none"> • need of external bias voltage • relatively low power density at small scale
Electromagnetic	<ul style="list-style-type: none"> • good for low frequencies (5-100Hz) • no external voltage source needed • suitable to drive low impedances 	<ul style="list-style-type: none"> • inefficient at MEMS scales: low magnetic field, micro-magnets manufacturing issues • large mass displacement required.

Main limits of resonant VEHs

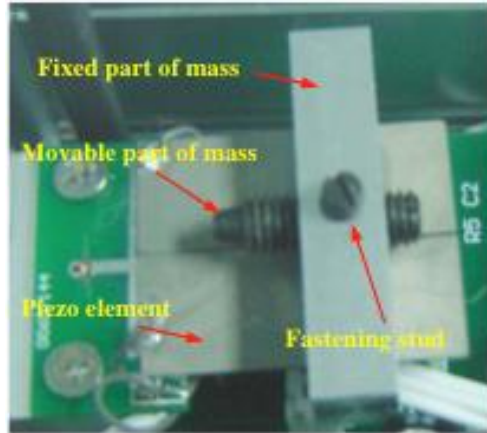
- narrow bandwidth that implies constrained resonant frequency-tuned applications
- Non-adaptation to variable vibration sources
- small inertial mass and high resonant frequency at micro/nano-scale -> most of vibration sources are below 100 Hz



At 20% off the resonance
the power falls by 80-90%

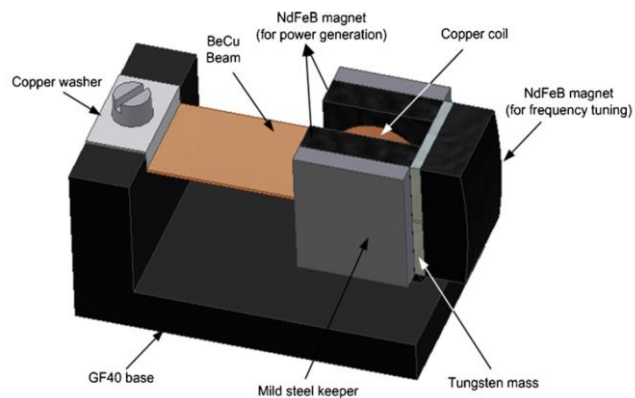
Beyond linear harvesting systems

Frequency tuning



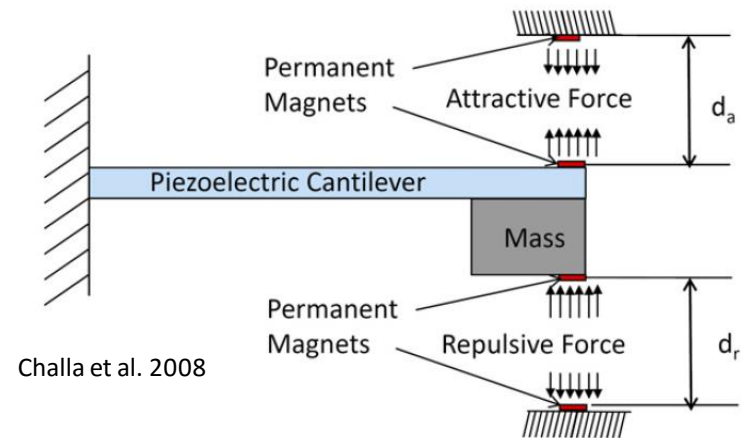
Piezoelectric cantilever with a movable mass

Wu et al. 2008

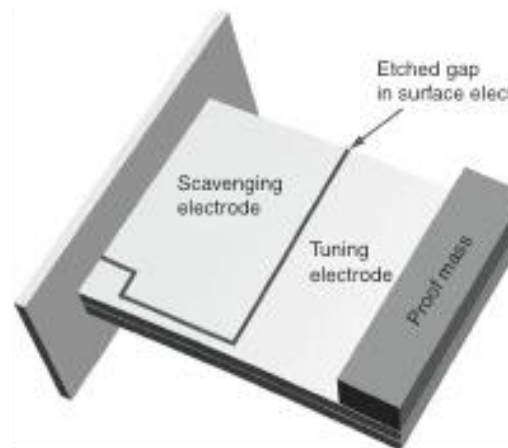


Zhu, et al. (2010). *Sensors and Actuators A: Physical*

Piezoelectric cantilever with magnetic tuning



Challa et al. 2008



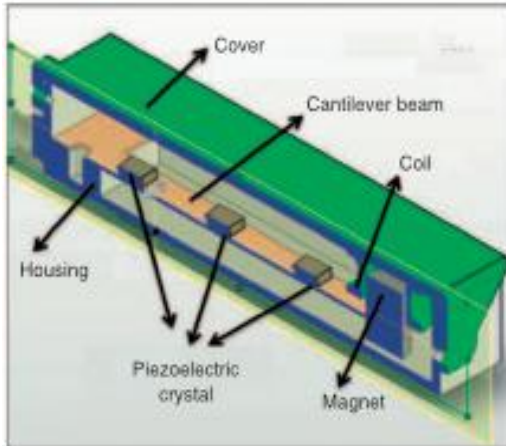
Piezoelectric beam with a scavenging and a tuning part

Roundy and Zhang 2004

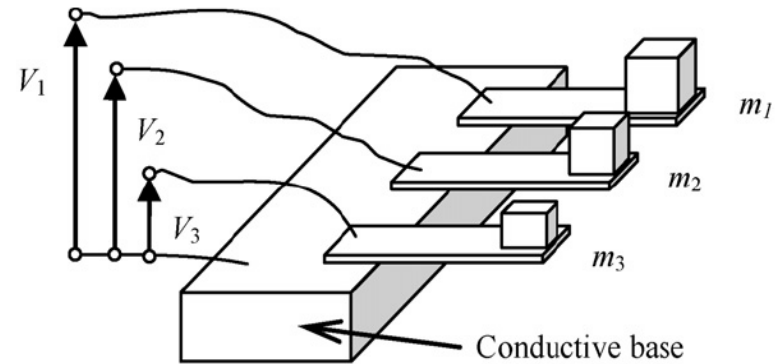
Beyond linear harvesting systems

Multimodal Energy Harvesting

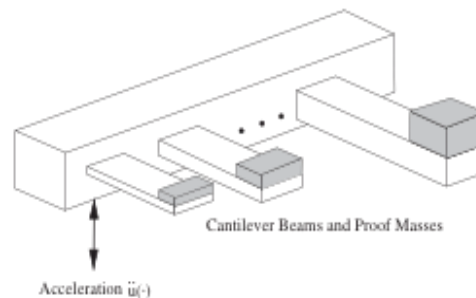
Tadesse et al. 2009



Hybrid harvester with piezoelectric and electromagnetic transduction mechanisms



Ferrari, M., et al. (2008). *Sensors and Actuators A: Physical*

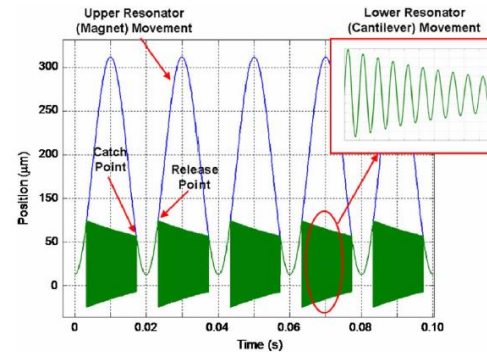
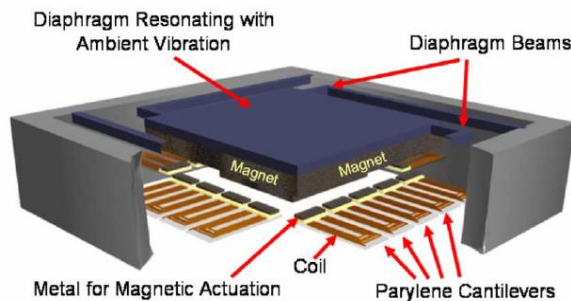


Piezoelectric cantilever arrays
with various lengths and tip masses

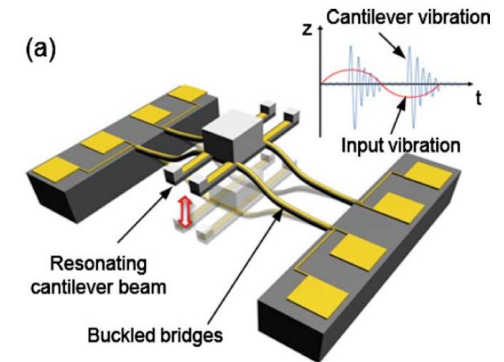
Shahruz 2006

Beyond linear harvesting systems

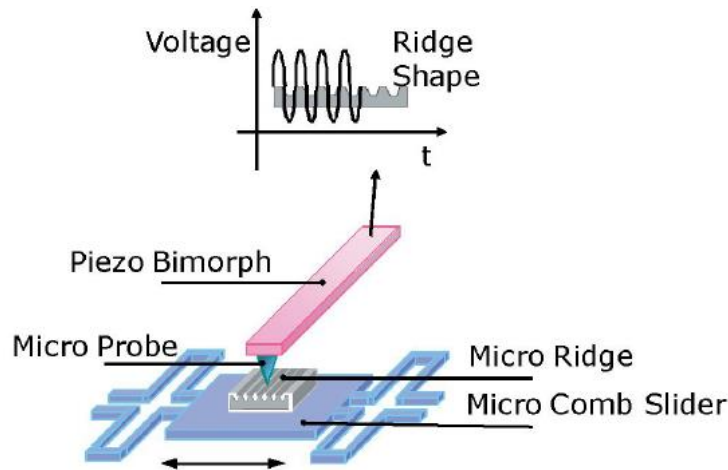
Frequency-up conversion



H. Kula and K. Najafi, IEEE Sensors Journal 8 (3), 261 (2008).

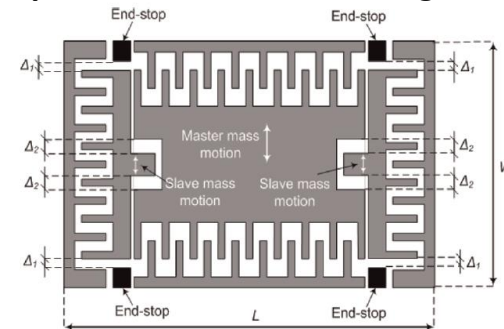


Jung, S.-M. et al. (2010). *Applied Physics Letters*



D.G. Lee et al. IEEE proc. (2007)

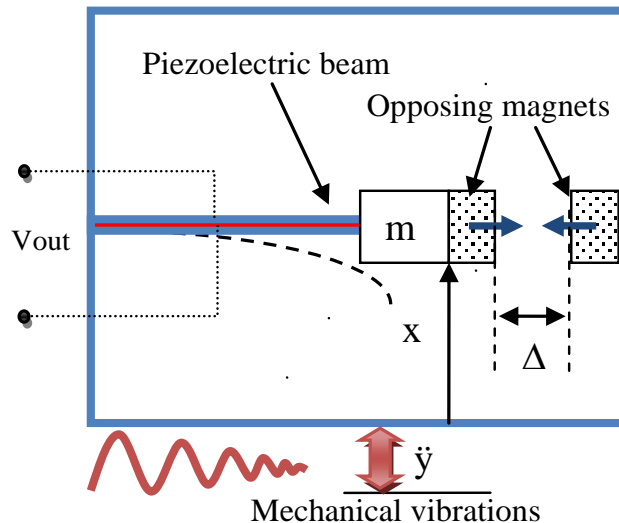
Impact electrostatic MEMS generator



Le, C. P., Halvorsen (2012). *Journal of Intelligent Material Systems and Structures*

Beyond linear harvesting systems

Nonlinear systems for vibration energy harvesting



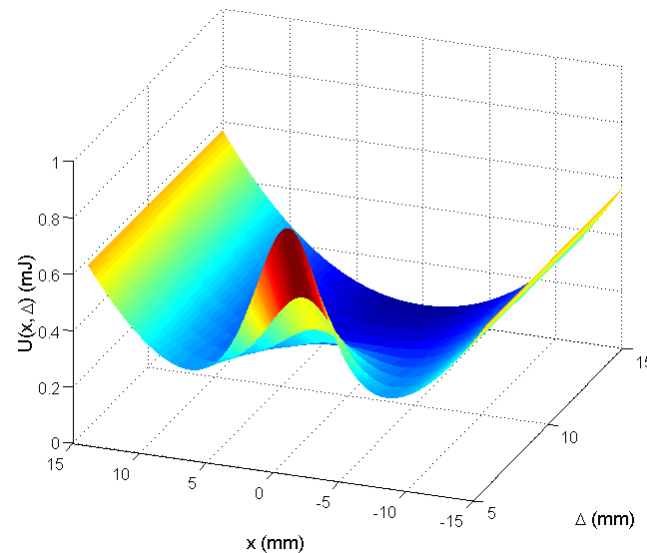
Magneto-elastic potential

Governing equations of a single-DOF piezo-magnetoelastic model

$$U(x, \Delta) = \frac{1}{2} K_{eff} x^2 + \frac{\mu_0}{2\pi} \frac{M_1 M_2}{(x^2 + \Delta^2)^{3/2}}$$

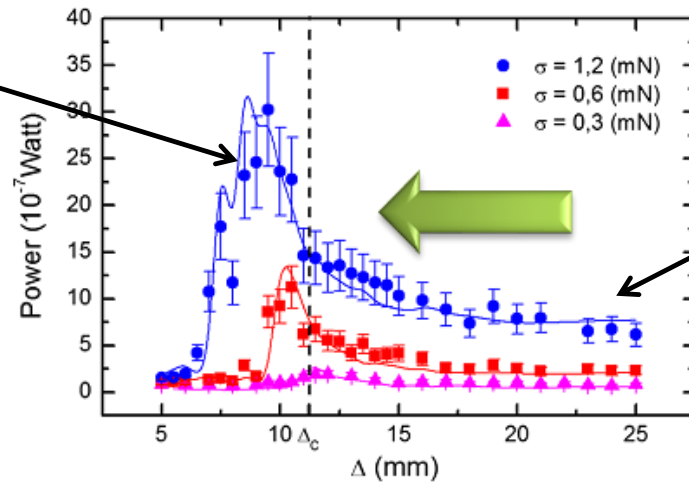
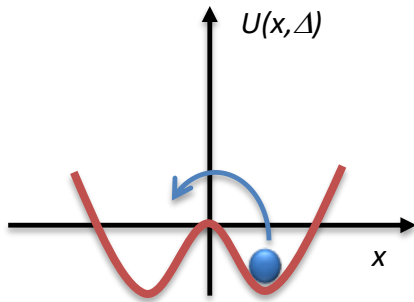
$$\begin{cases} m\ddot{x}(t) + \delta\dot{x}(t) + K_{eff}x(t) + \frac{\partial U(x, \Delta)}{\partial x} + K_v V(t) = -m\ddot{y}(t) \\ \dot{V}(t) + \frac{1}{\tau} V(t) = K_e \dot{x}(t); & \tau = R_L C_p \end{cases}$$

Cottone, F., H. Vocca & L. Gammaitoni. *PRL*, 102 (2009).



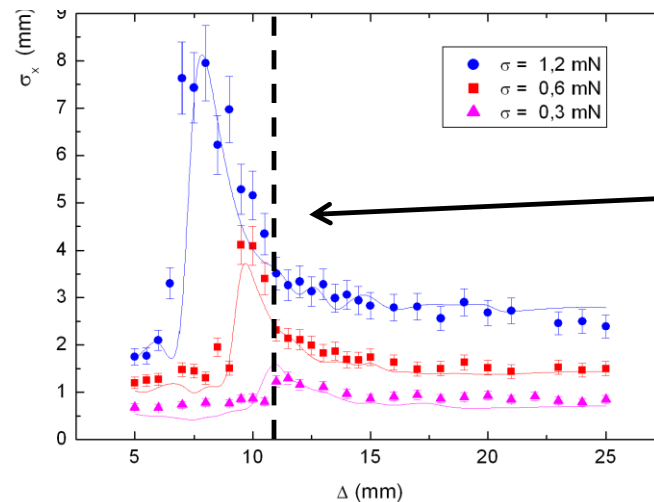
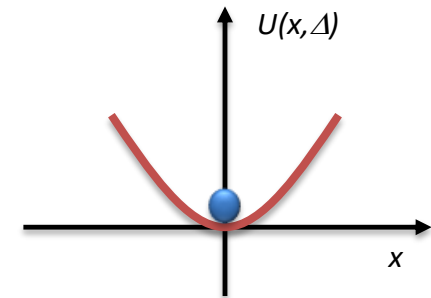
Beyond linear harvesting systems

Bistable: inter-well and intra-well oscillations



Resonant monostable

$\Delta = 25$ mm



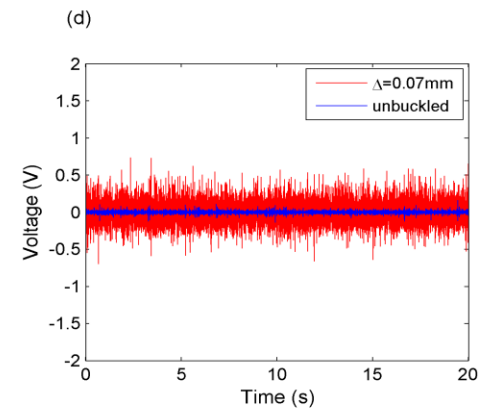
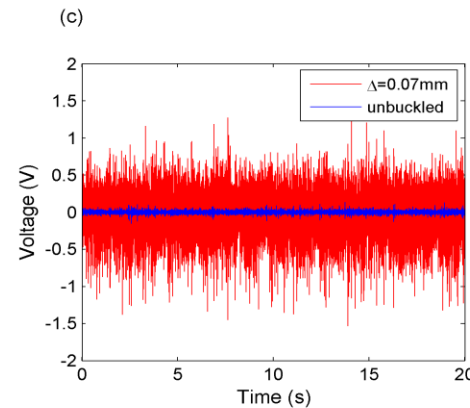
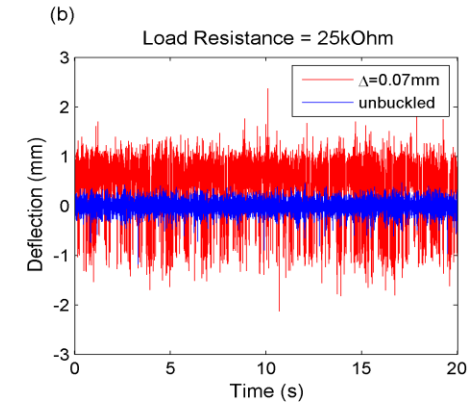
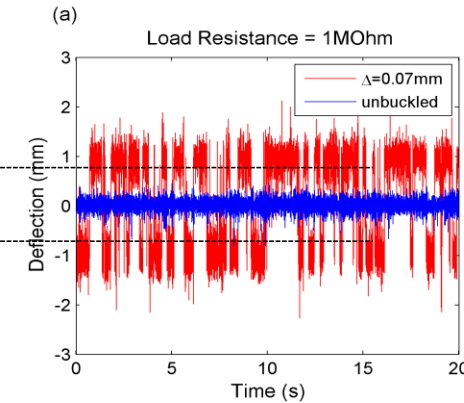
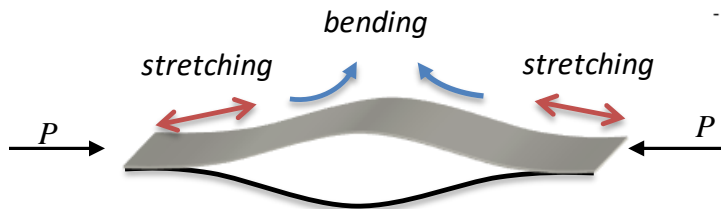
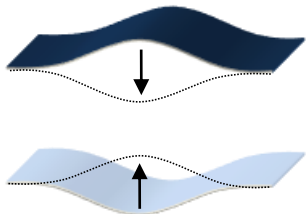
Bifurcation point

Beyond linear harvesting systems

Buckled beam piezoelectric harvesters

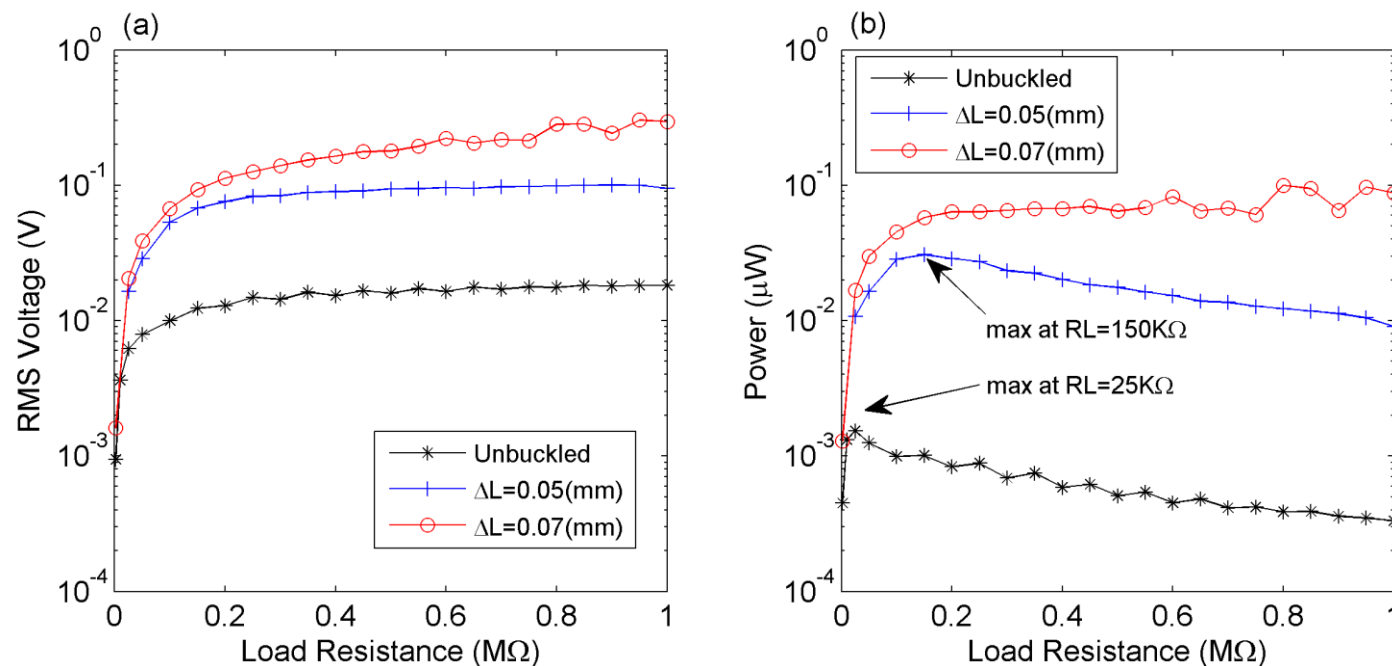
Cottone, F., Gammaitoni, L., Vocca, H., Ferrari, M., & Ferrari, V. (2012). *Smart materials and structures*, 21(3), 035021

Snapping between buckled states



Beyond linear harvesting systems

Experimental and numerical results



Cottone, F., L. Gammaitoni, H. Vocca, M. Ferrari & V. Ferrari (2012) Smart materials and structures, 21, 2012.

Beyond linear harvesting systems

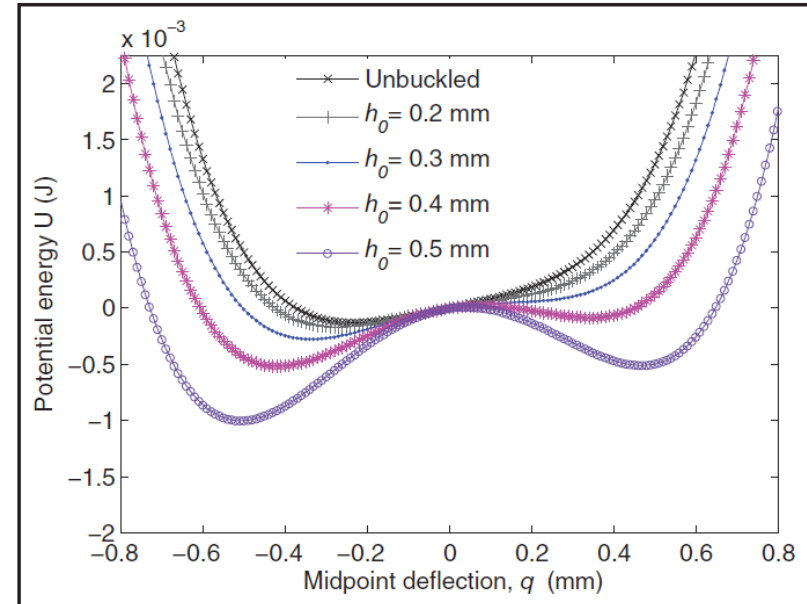
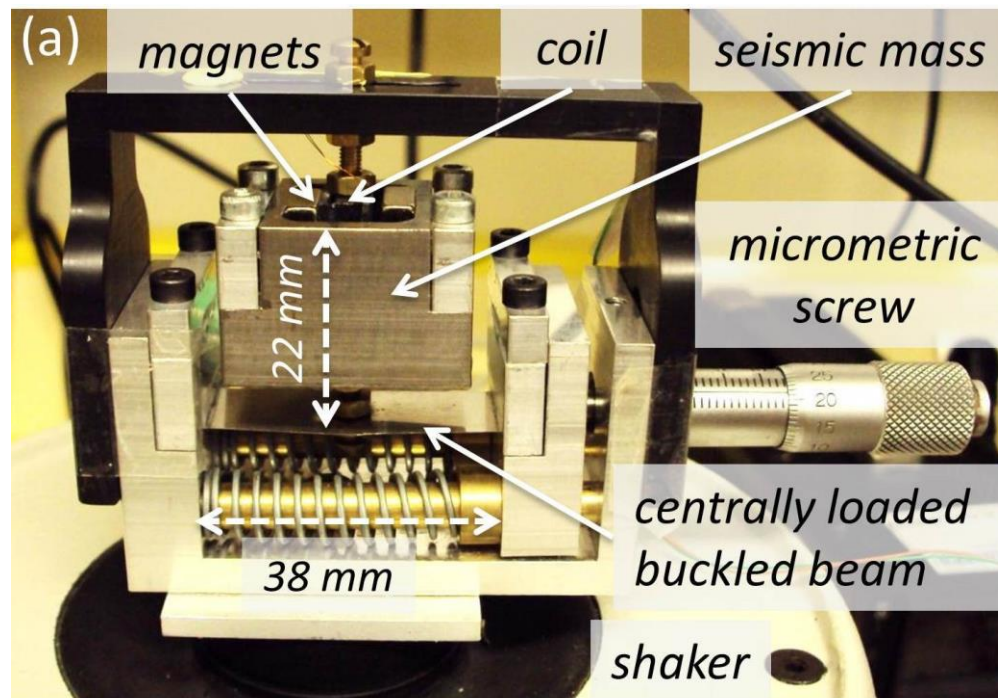


Figure 3. Potential energy of the system for increasing values of buckling height h_0 .

F. Cottone et al. *J. Intell. Mater. Syst. Struct.* 2014.

Beyond linear harvesting systems

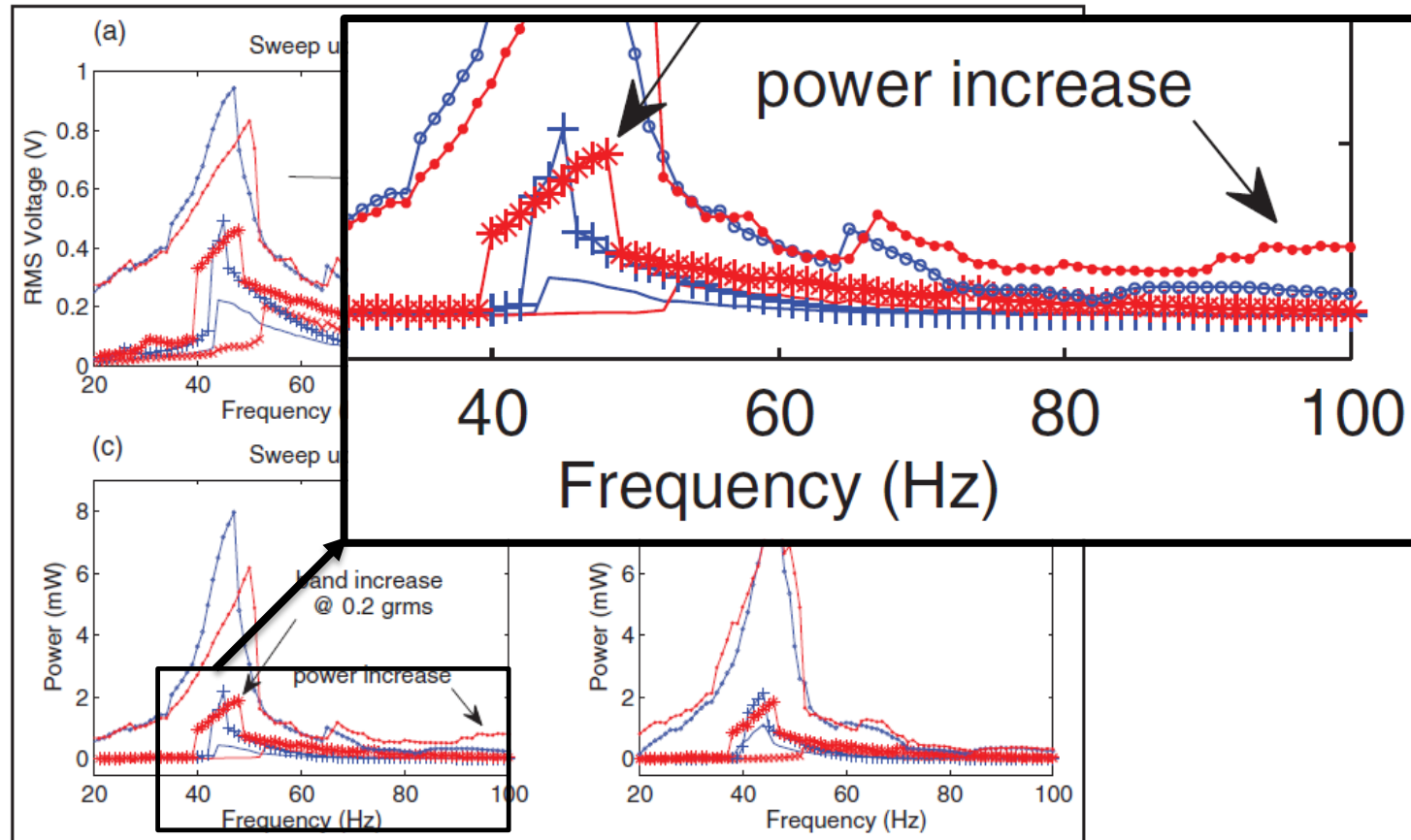


Figure 6. Experimental comparison of unbuckled- and buckled-beam ($h_0 = 0.3$ mm) generators for up (left column) and down (right column) frequency sweeps with acceleration amplitudes of 0.1, 0.2, and 0.5 g_{rms} . (a and b) rms voltage and (c and d) the corresponding power dissipated across the optimal load resistance $R_L = 112 \Omega$.

rms: root mean square.

Figure of Merit

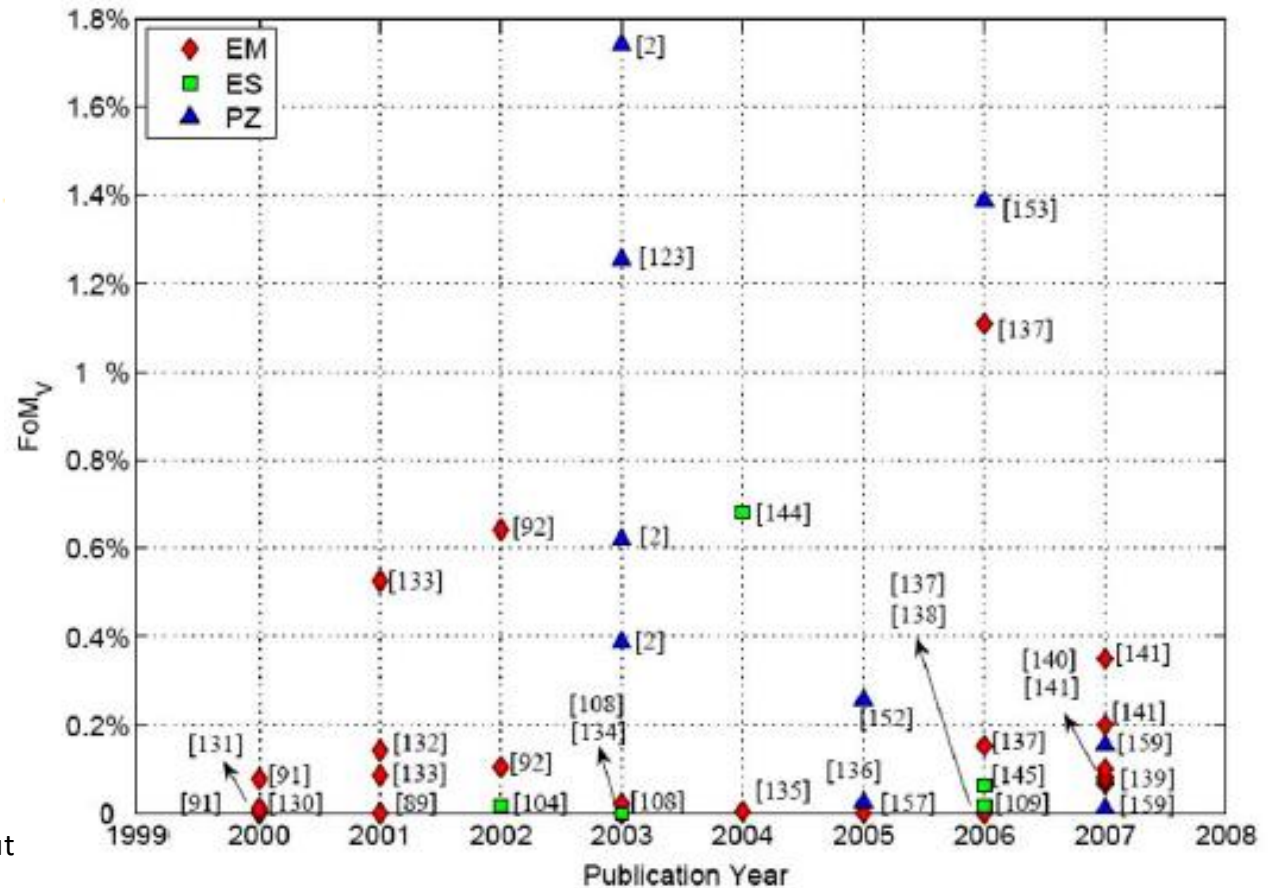
Volume figure of merit

$$\text{FoM}_V = \frac{\text{Useful Power Output}}{\frac{1}{16} Y_0 \rho_{Au} Vol^{\frac{4}{3}} \omega^3}$$

Bandwidth figure of merit

$$\text{FoM}_{BW} = \text{FoM}_V \times \frac{\delta\omega_1 \text{ dB}}{\omega}$$

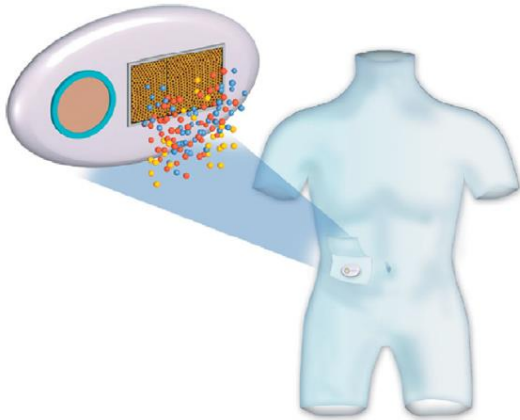
Frequency range within which the output power is less than 1 dB below its maximum value



Mitcheson, P. D., E. M. Yeatman, et al. (2008). *Proceedings of the IEEE* **96**(9): 1457-1486.

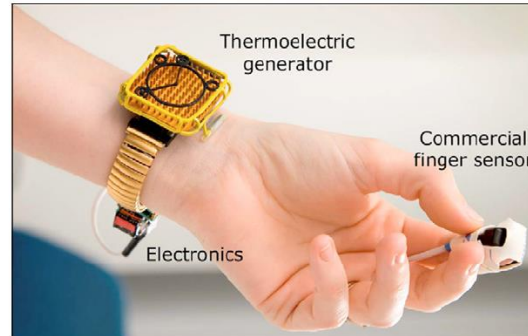
Microscale energy harvesters

MEMS-based drug delivery systems



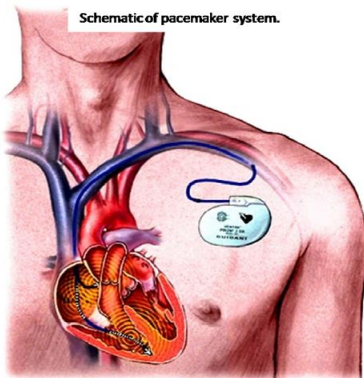
Bohm S. et al. 2000

Body-powered oximeter



Leonov, V., & Vullers, R. J. (2009).

Heart powered pacemaker



Pacemaker consumption is **40uW**.

Beating heart could produce **200uW** of power

D. Tran, Stanford Univ. 2007

Micro-robot for remote monitoring

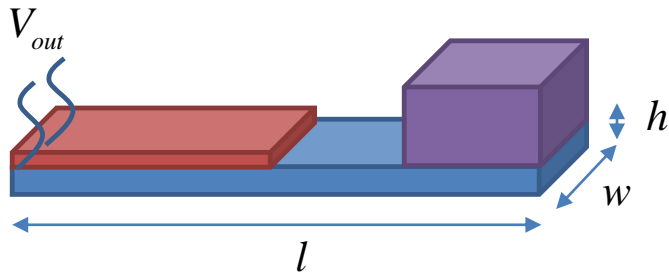


The input power a 20 mg robotic fly is **10 – 100 uW**

A. Freitas Jr., Nanomedicine, Landes Bioscience, 1999

Microscale energy harvesters: scaling issues

First order power calculus with William and Yates model



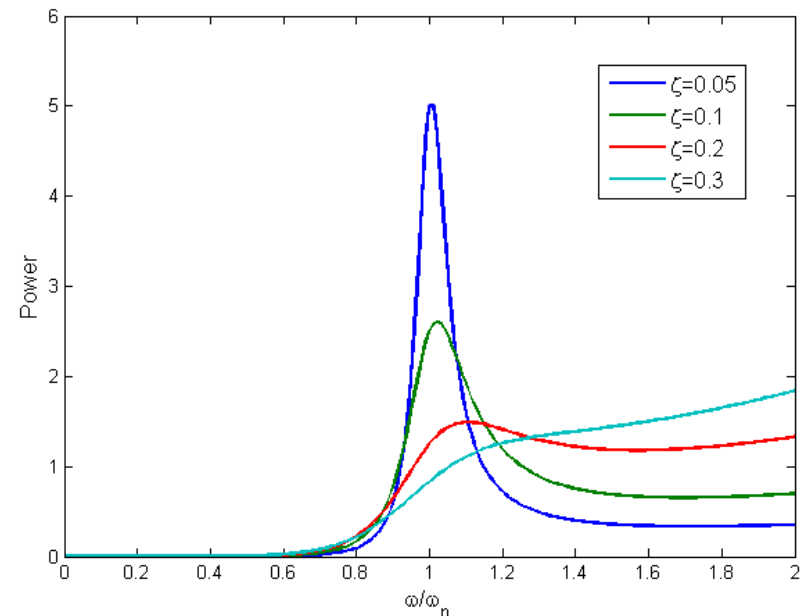
$$\omega_n = 2\pi C_n \sqrt{\frac{E}{\rho}} \frac{h}{l^2}$$

$$k = \xi \frac{Ewh^3}{l^3}$$

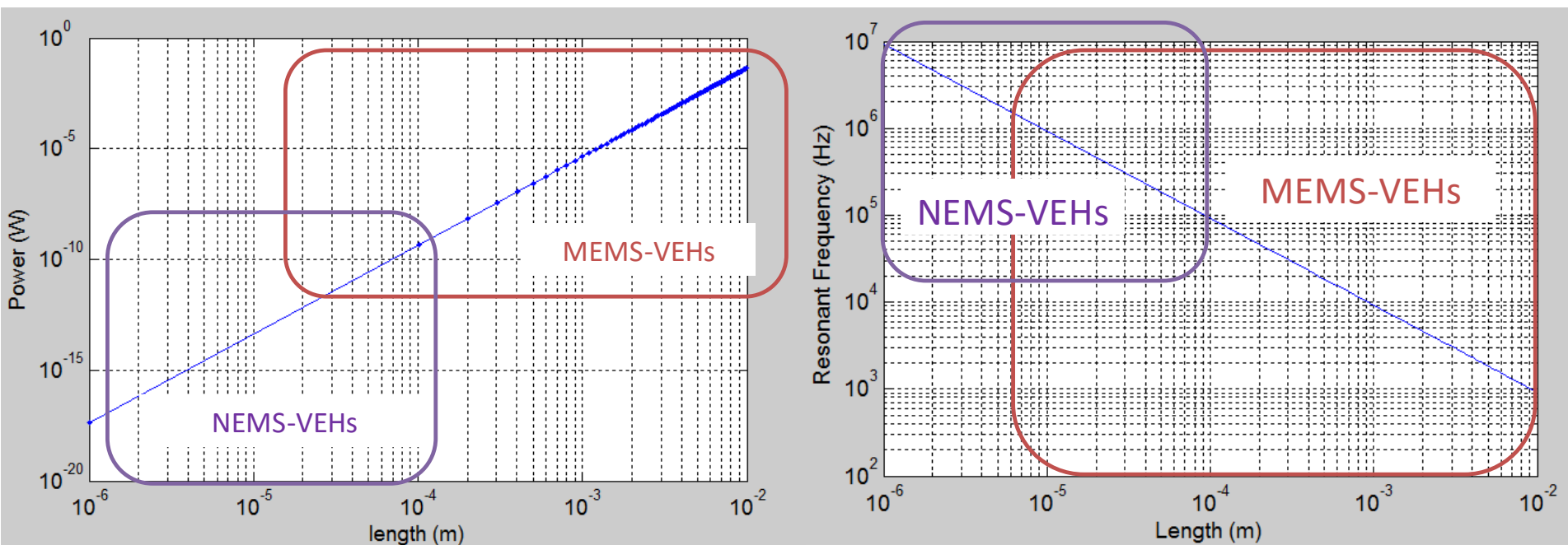
Boundary conditions	C1
doubly clamped	1,03
cantilever	0,162

Boundary conditions	Uniform load ξ	Point load ξ
doubly clamped	32	16
cantilever	0,67	0,25

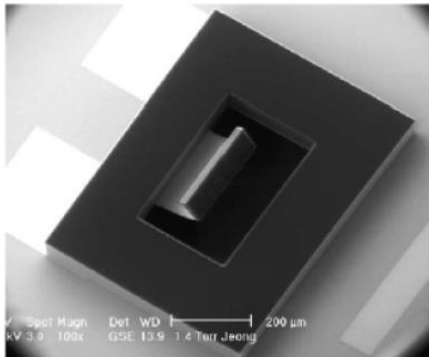
- Low efficiency off resonance
- High resonant frequency at miniature scales
- **Power** $\rightarrow A^2/l^4$ where A is the acceleration and l the linear dimension



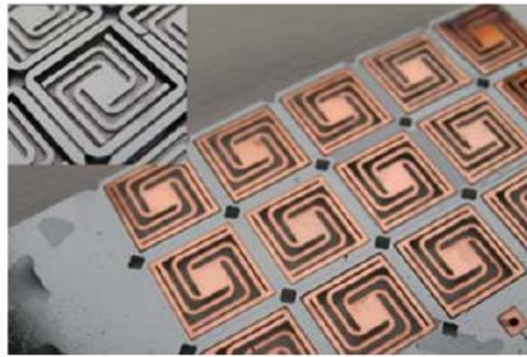
Microscale energy harvesters: scaling issues



Microscale energy harvesters: scaling issues



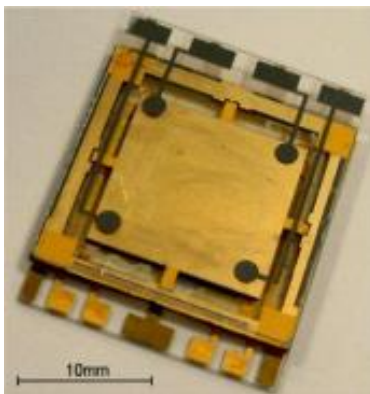
Jeon et al. 2005



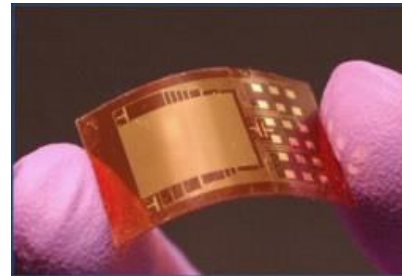
EM generator, Miao et al. 2006



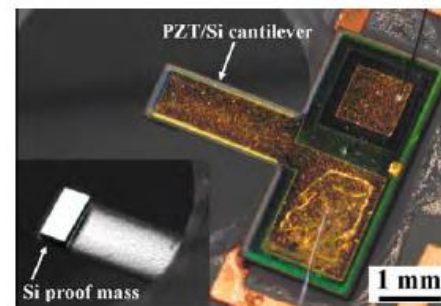
Chang, MIT 2013



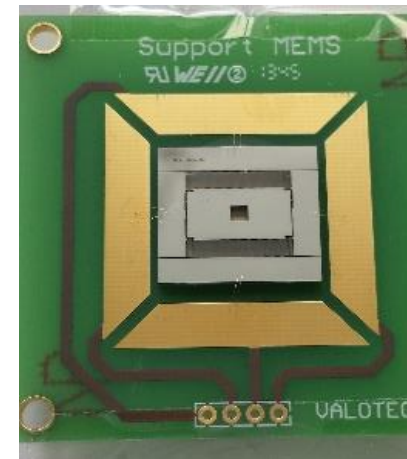
Mitcheson 2005 (UK)
Electrostatic generator 20Hz
2.5uW @ 1g



ZnO nanowires
Wang, Georgia Tech (2005)



D. Briand, EPFL 2010



Cottone F., Basset P. ESIEE Paris 2013

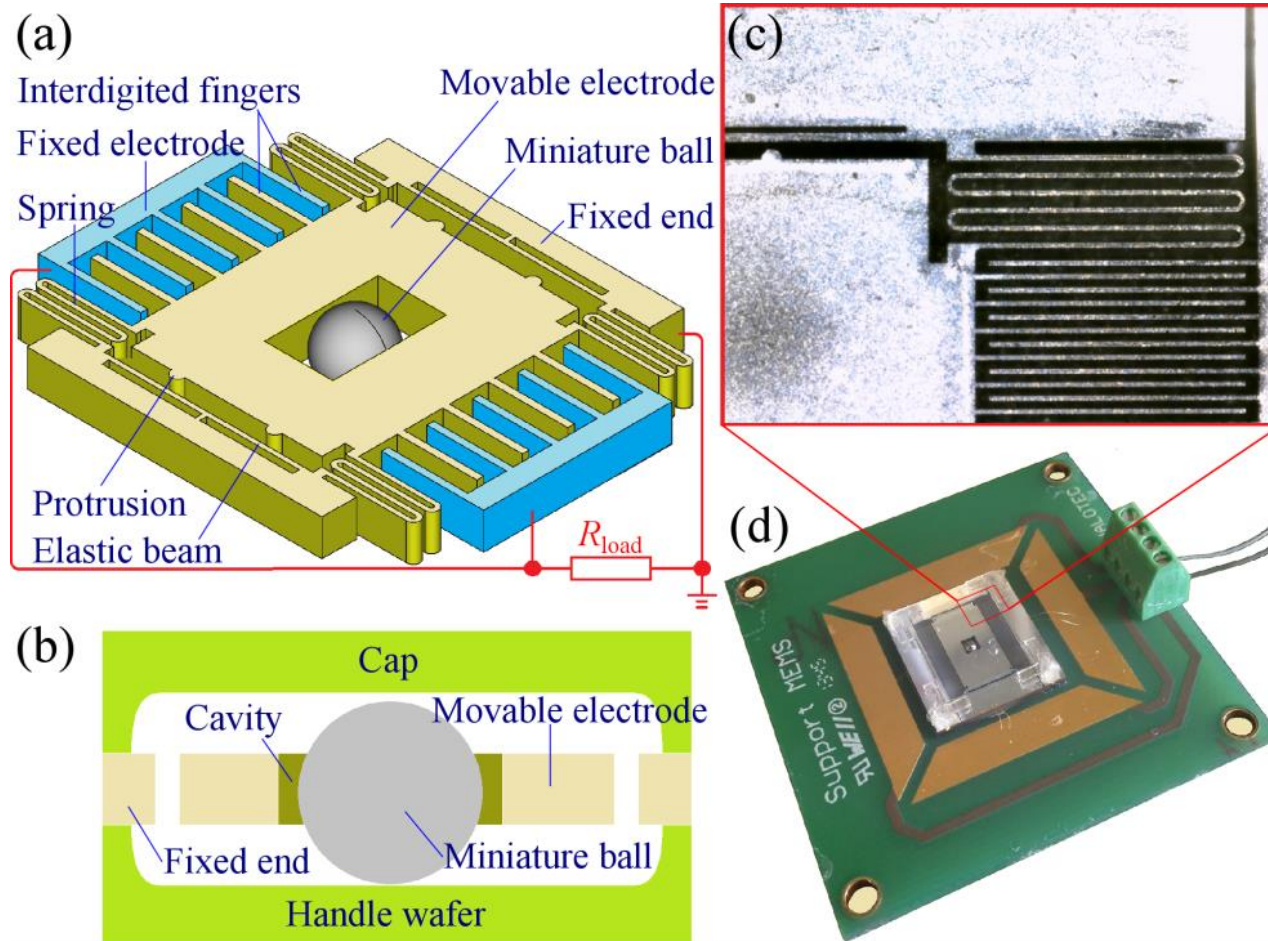


2005

2015

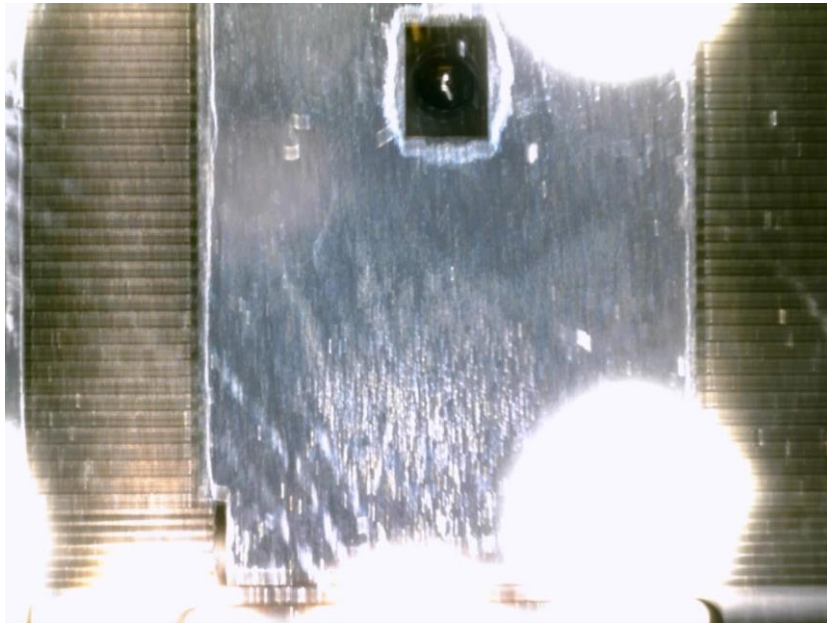
Low-frequency MEMS electrostatic VEH

Prototype fabrication process

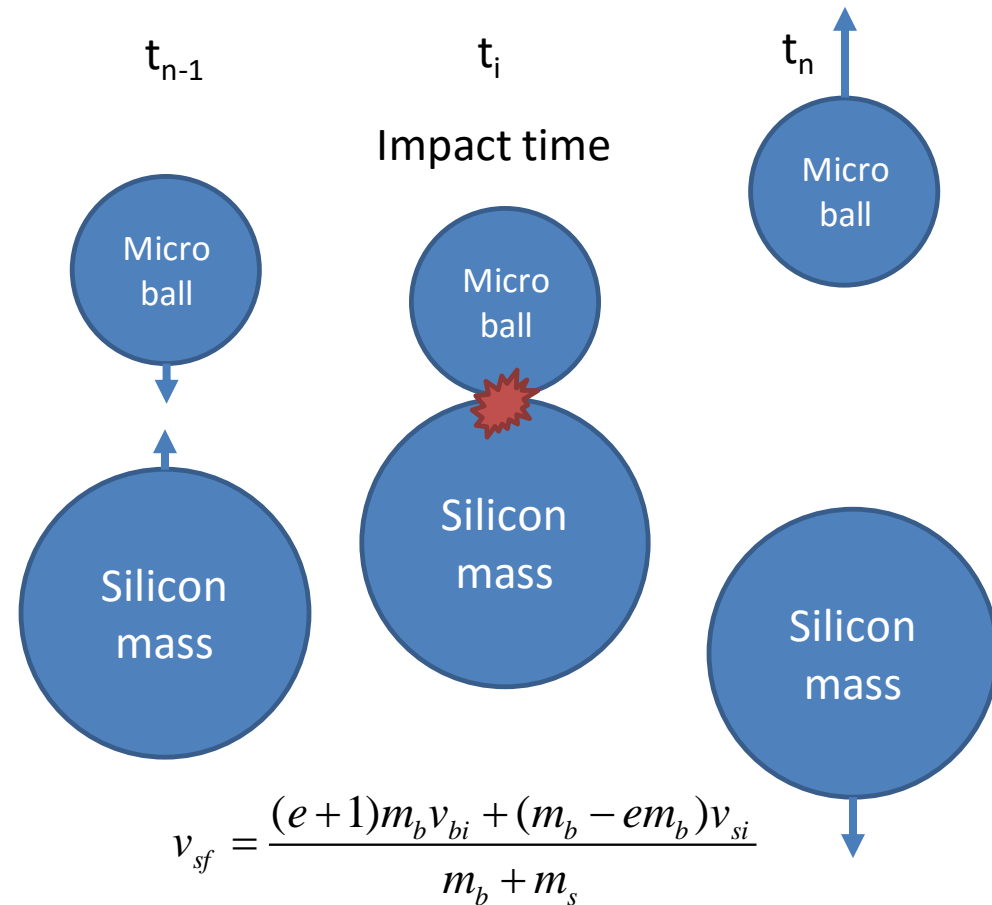


Low-frequency MEMS electrostatic VEH

Experimental test



Working principle

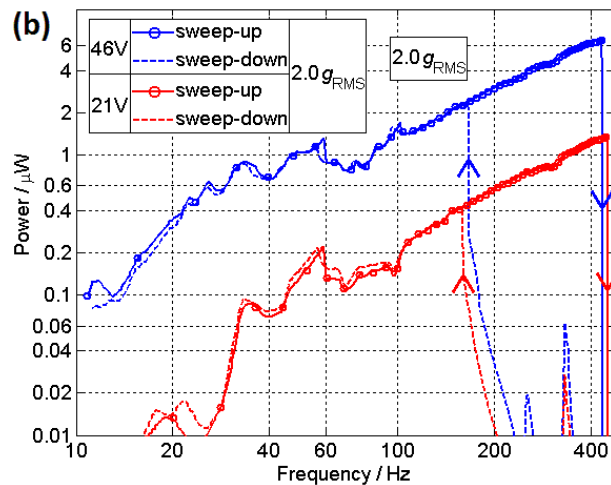
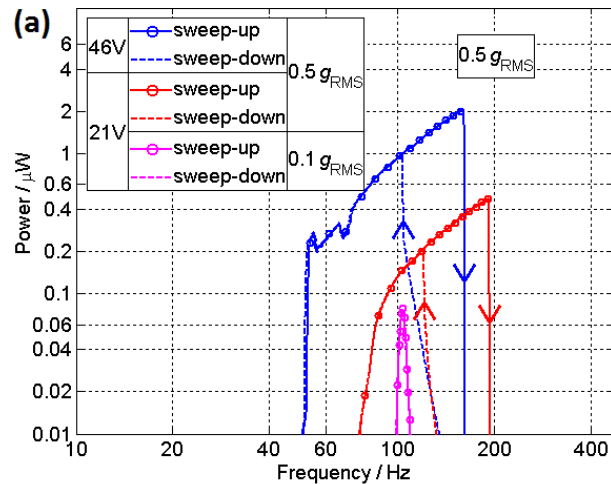


Velocity Amplified Energy Harvester

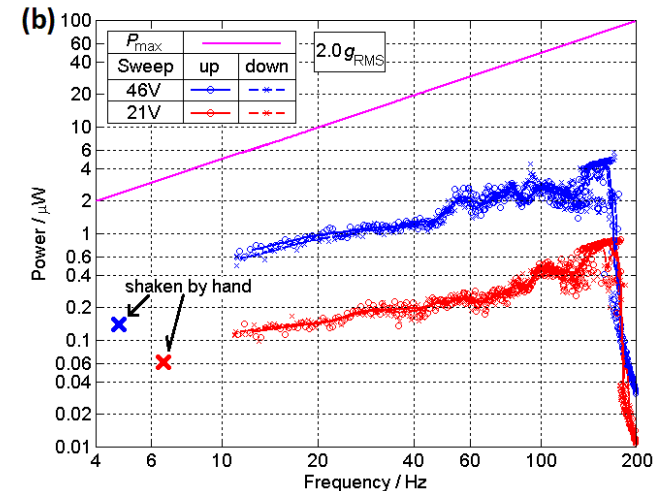
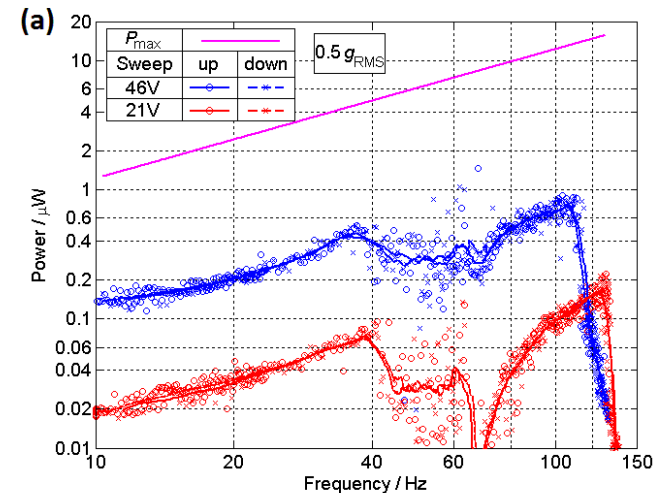
At Stoke Institute, University of Limerick, Ireland

Low-frequency MEMS electrostatic VEH

without micro-ball

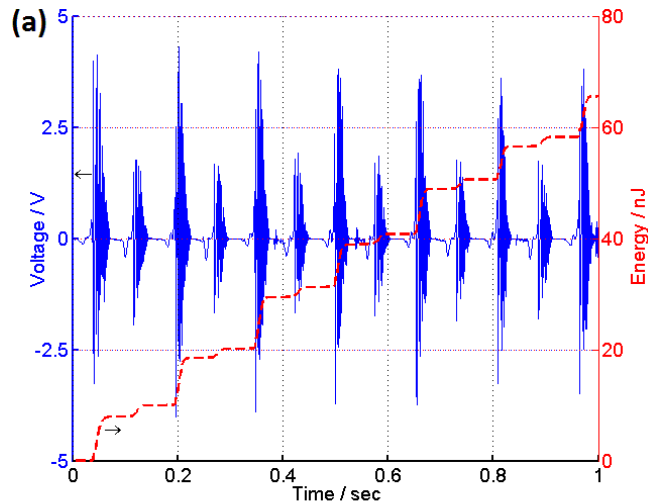


with micro-ball



Y. Lu, F. Cottone, S. Boisseau, F. Marty, D. Galayko, and P. Basset, Appl. Phys. Lett. 2015.

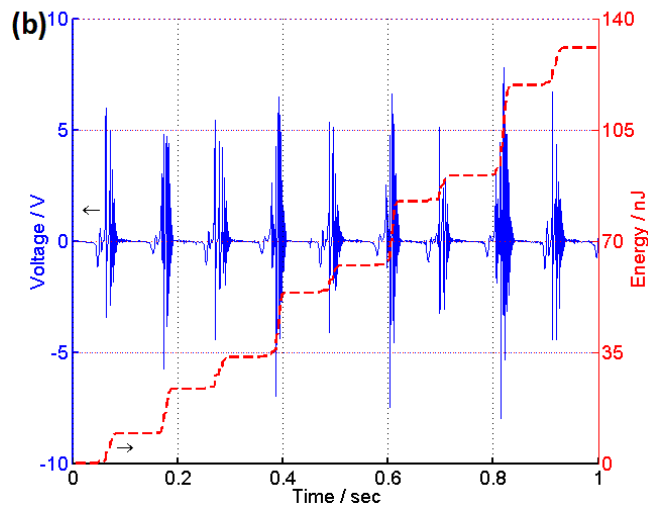
Low-frequency MEMS electrostatic VEH



TEST with hand shaking of the transient output voltage and extracted energy.

(a) $V_{bias}=21$ V, $a=2.0$ grms, $f=6.5$ Hz;

(b) $V_{bias}=46$ V, $a=2.0$ grms, $f=4.7$ Hz



A 47- μ F capacitor has been also charged through a bridge diode rectifier to 3.5 V to supply a wireless temperature sensor node.

Performance comparison

Vibration type	MEMS Direction	Accel. (gRMS)	Main input Freq. (Hz)	Vbias (V)	Power (uW)	Power Density (uW/cm ³)
Man walking	X	0.39	4.15	20	1.34	13.40
Man walking	Y	0.27	2.1	20	0.793	7.93
Man walking	Z	0.41	2.44	20	1.15	11.50
Man running	Z	1.20	3.3	20	14.9	142.00

Table 2 Comparison of Effectiveness of Published Electrostatic Motion Harvesters

Author	Reference	Generator Volume [cm ³]	Proof Mass [g]	Input Amplitude [μ m]	Input Frequency [Hz]	Z_I [μ m]	Power (un-processed) [μ W]	Power (pro-cessed) [μ W]	Power Density [μ W/cm ³]	Harvester Effec-tiveness [%]	Volume Figure of Merit [%]
Tashiro	[104]		640	380	4.76	19000		58		0.09	
Tashiro	[142]	15	780	9000	6		36		2.42		0.02
Mizuno	[108]	0.6	0.7	0.64	743	4.9	7.4×10^{-6}		1.23×10^{-3}	6.6×10^{-6}	1.86×10^{-9}
Miyazaki	[143]		5	1	45	30		0.21		12.4	
Arakawa	[144]	0.4	0.65	1000	10	1000	6		15	7.42	0.68
Despesse	[145]	18	104	90	50	90	1760	1000	56	7.66	0.06
Yen	[146]				1500			1.8			
Tsutsumino	[147]			600	20	600	278				
Tsutsumino	[148]			1000	20	1000	6.4				
Mitcheson	[109]	0.6	0.12	1130	20	100	2.4		4	17.9	0.02

Almost 1 order of magnitude higher than average power density of previous works

P. D. Mitcheson, et al, *Proceedings of the IEEE*, vol. 96, pp. 1457-1486, 2008.

Final considerations

- **Energy harvesting systems** will enable low-power Technology to be **completely autonomous**
- **Energy harvesting systems** can be improved by:
 - **Nonlinear dynamic**: bistable systems, frequency-up converters, impacting masses, electrostatic softening
 - Innovative **electro-active materials** (electrets, lead-free piezo)
 - Miniaturization: research on innovative **nanostructured piezoelectric, magnetic** and **electrets** materials, frequency-up conversion
- There is need for improvement for **low-power electronics**
 - Low-consumption components
 - Efficient conditioning circuit
 - Power-aware software